

# Circuit Theory

## Chapter 14

### Frequency Response

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# Frequency Response

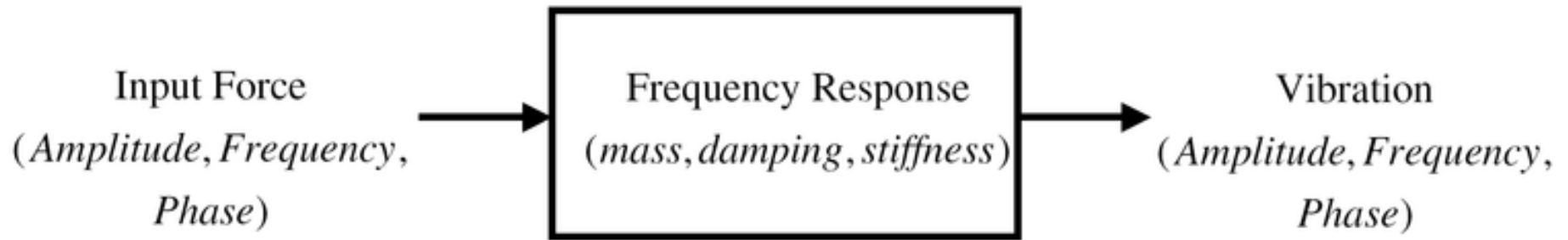
## Chapter 14

- 14.1 Introduction
- 14.2 Transfer Function
- 14.3 Series Resonance
- 14.4 Parallel Resonance
- 14.5 Passive Filters

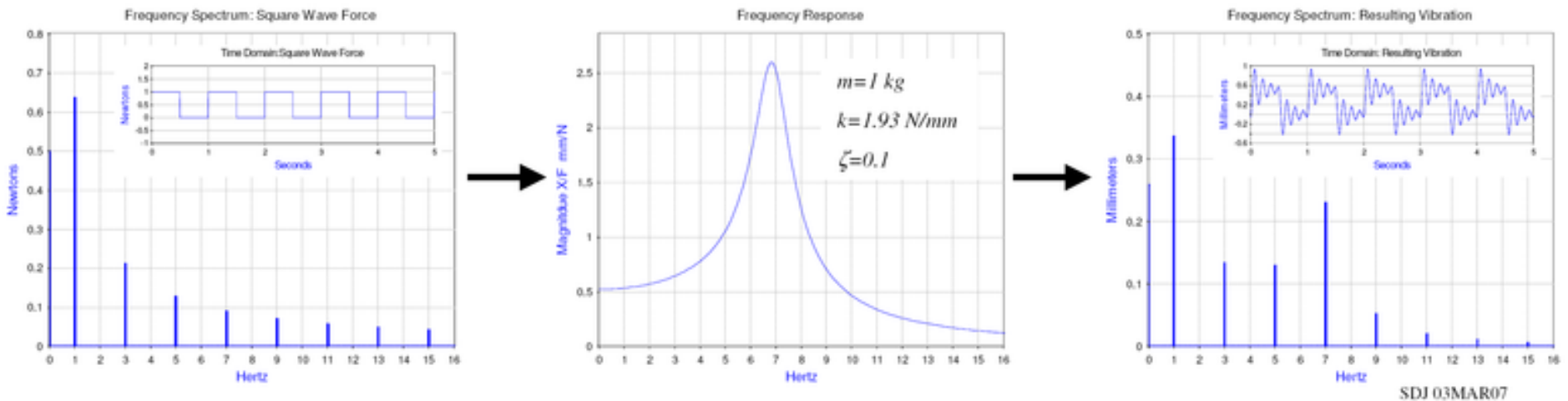
# 14.1 Introduction (1)

What is FrequencyResponse of a Circuit?

**It is the variation in a circuit's behavior with change in signal frequency and may also be considered as the variation of the gain and phase with frequency.**



$$F(\omega) \times H(\omega) = X(\omega)$$

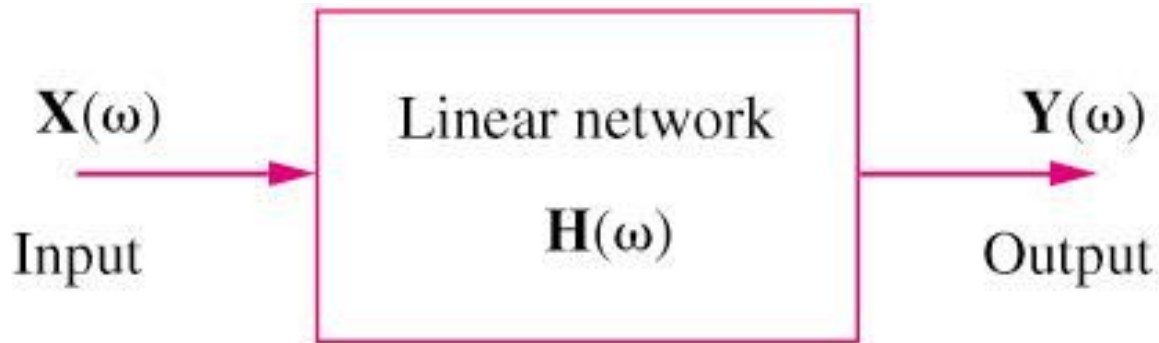


[http://en.wikipedia.org/wiki/Tacoma\\_Narrows\\_Bridge\\_\(1940\)](http://en.wikipedia.org/wiki/Tacoma_Narrows_Bridge_(1940))

<http://lpsa.swarthmore.edu/Analog/ElectricalMechanicalAnalog.html>

# 14.2 Transfer Function (1)

- The transfer function  $H(\omega)$  of a circuit is the **frequency-dependent ratio** of a phasor output  $\underline{Y}(\omega)$  (an element voltage or current ) to a phasor input  $\underline{X}(\omega)$  (source voltage or current).



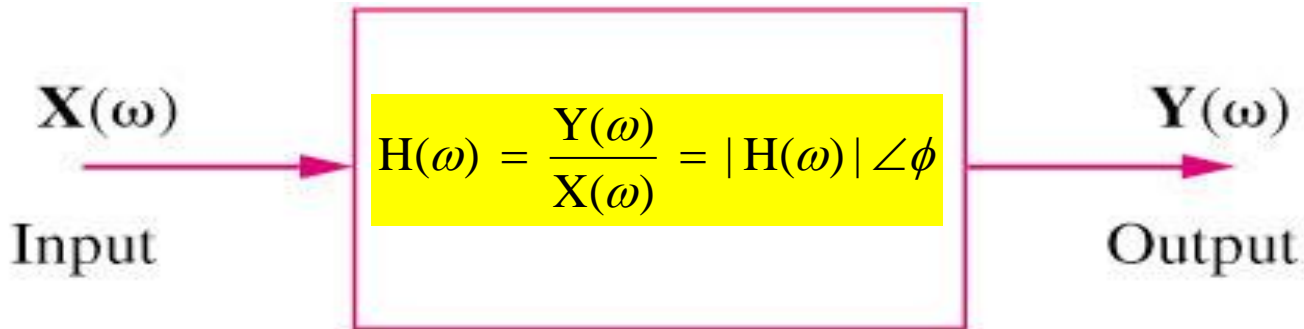
$$H(\omega) = \frac{Y(\omega)}{X(\omega)} = |H(\omega)| \angle \phi$$

# 14.2 Transfer Function (2)

- Four possible transfer functions:

$$H(\omega) = \text{Voltage gain} = \frac{V_o(\omega)}{V_i(\omega)}$$

$$H(\omega) = \text{Transfer Impedance} = \frac{V_o(\omega)}{I_i(\omega)}$$



$$H(\omega) = \text{Current gain} = \frac{I_o(\omega)}{I_i(\omega)}$$

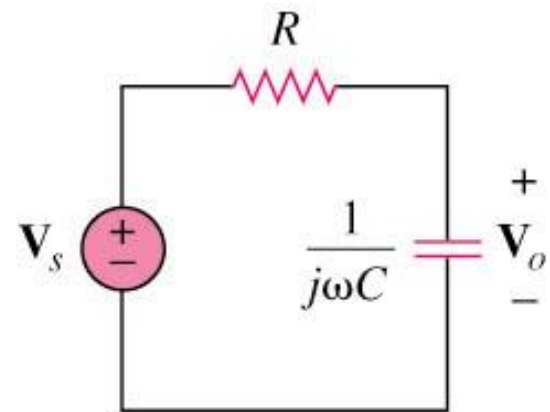
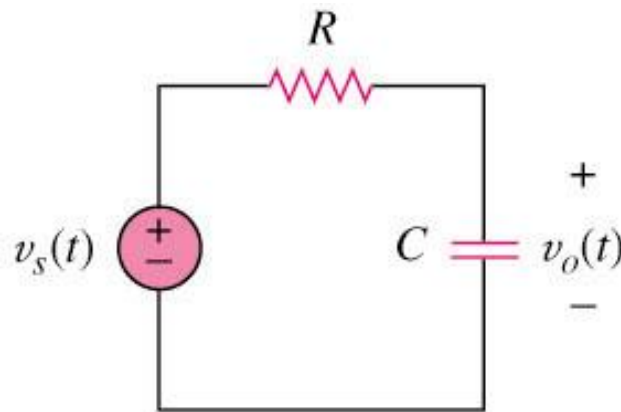
$$H(\omega) = \text{Transfer Admittance} = \frac{I_o(\omega)}{V_i(\omega)}$$

# Example 1

For the RC circuit shown below, obtain the transfer function  $V_o/V_s$  and its frequency response. Let  $v_s = V_m \cos \omega t$ .

Voltage Divider

$$V_o = \left( \frac{\cancel{1/j\omega C}}{R + \frac{1}{j\omega C}} \right) V_s$$



$$H(\omega) = \frac{V_o}{V_s} = \frac{1}{1 + j\omega RC} \quad \text{(a)}$$

let  $\omega_0 = \frac{1}{RC}$

$$H(\omega) = \frac{1}{1 + j \frac{\omega}{\omega_0}} \quad \text{(b)}$$

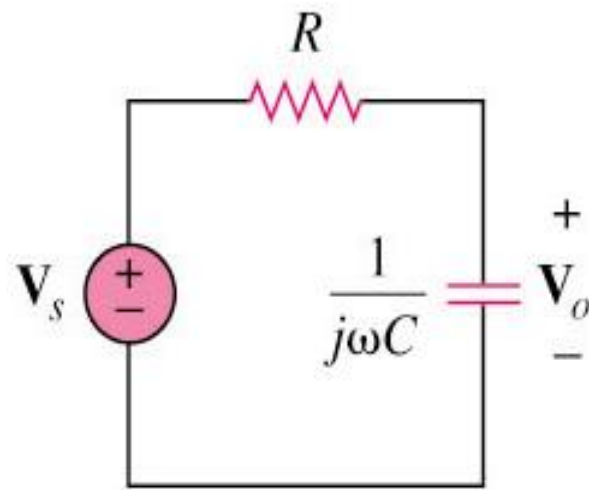
Polar Form

$$H(\omega) = \frac{1}{\sqrt{1 + \left(\frac{\omega}{\omega_0}\right)^2}} \angle \tan^{-1}\left(\frac{\omega}{\omega_0}\right)$$

$$= \frac{1}{\sqrt{1 + \left(\frac{\omega}{\omega_0}\right)^2}} \angle -\tan^{-1}\left(\frac{\omega}{\omega_0}\right)$$

magnitude      Phase angle

$\omega = 0$   
 $H(0) = 1 \angle 0^\circ$   
 $\omega = \infty$   
 $H(\infty) = 0 \angle -90^\circ$   
 $\omega = \omega_0$   
 $H(\omega_0) = 0.707 \angle -45^\circ$   
Low Pass Filter



## Solution:

The transfer function is

$$H(\omega) = \frac{V_o}{V_s} = \frac{\frac{1}{j\omega C}}{R + \frac{1}{j\omega C}} = \frac{1}{1 + j\omega RC}$$

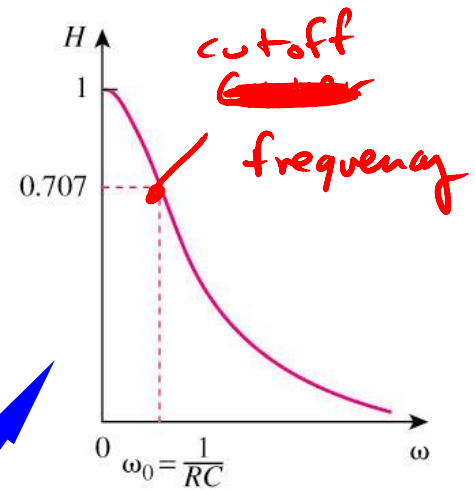
The magnitude is

$$|H(\omega)| = \frac{1}{\sqrt{1 + (\omega/\omega_o)^2}}$$

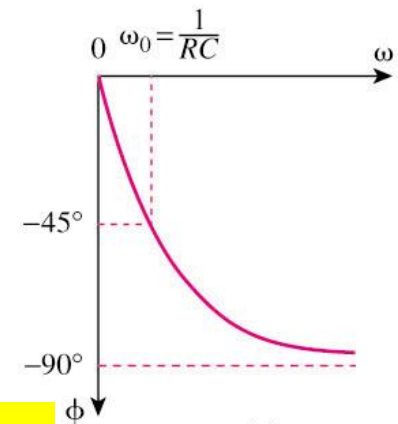
The phase is

$$\phi = -\tan^{-1} \frac{\omega}{\omega_o}$$

$$\omega_o = 1/RC$$



(a)

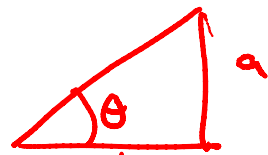


(b)

**Low Pass Filter**



## Example 2



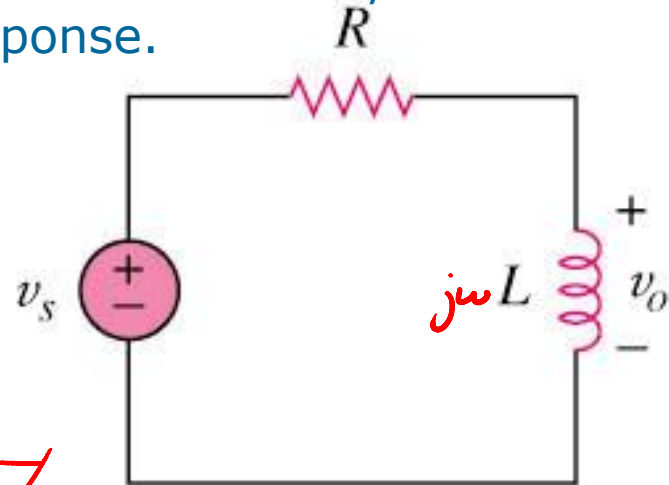
$$\tan^{-1}\left(\frac{a}{b}\right) = 90 - \tan^{-1}\left(\frac{b}{a}\right)$$

Obtain the transfer function  $V_o/V_s$  of the RL circuit shown below, assuming  $v_s = V_m \cos \omega t$ . Sketch its frequency response.

$$V_o = \frac{j\omega L}{R + j\omega L} V_s$$

$$H(\omega) = \frac{V_o}{V_s} = \frac{j\omega L / j\omega L}{(R + j\omega L) / j\omega L} = \frac{1}{1 + \frac{R}{j\omega L}}$$

let  $\omega_0 = \frac{R}{L}$



$$\rightarrow H(\omega) = \frac{1}{1 - \frac{j\omega_0}{\omega}} \xrightarrow{\text{Polar Form}} = \frac{1}{\sqrt{1 + \left(\frac{\omega_0}{\omega}\right)^2}} \angle -\tan^{-1}\left(\frac{\omega_0}{\omega}\right) = \frac{1}{\sqrt{1 + \left(\frac{\omega_0}{\omega}\right)^2}} \angle \tan^{-1}\left(\frac{\omega}{\omega_0}\right)$$

$$= \left( \frac{1}{\sqrt{1 + \left(\frac{\omega_0}{\omega}\right)^2}} \right) \angle 90 - \tan^{-1}\left(\frac{\omega}{\omega_0}\right)$$

$$H(0) = 0 \angle 90$$

$$H(\omega_0) = .707 \angle 45$$

$$H(\infty) = 1 \angle 0$$

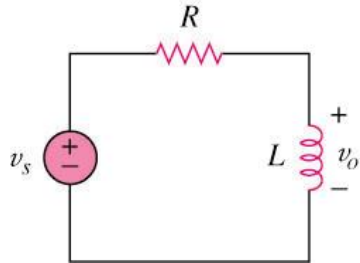
HIGH PASS  
FILTER

# 14.2 Transfer Function (6)

## Solution:

The transfer function is

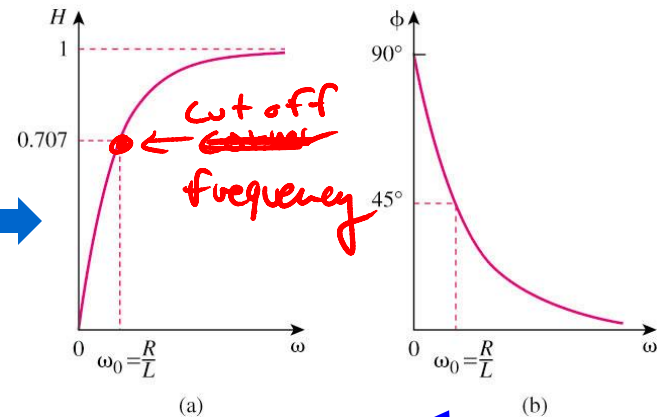
$$H(\omega) = \frac{V_o}{V_s} = \frac{j\omega L}{R + j\omega L} = \frac{1}{1 + \frac{R}{j\omega L}}$$



The magnitude is

$$H(\omega) = \frac{1}{\sqrt{1 + \left(\frac{\omega_o}{\omega}\right)^2}}$$

High Pass Filter



The phase is

$$\phi = \angle 90^\circ - \tan^{-1} \frac{\omega}{\omega_o}$$

$$\omega_o = R/L$$

Ex3) Find the transfer function  $H(\omega) = V_{out}/V_{in}$ .  
What type of filter is it?

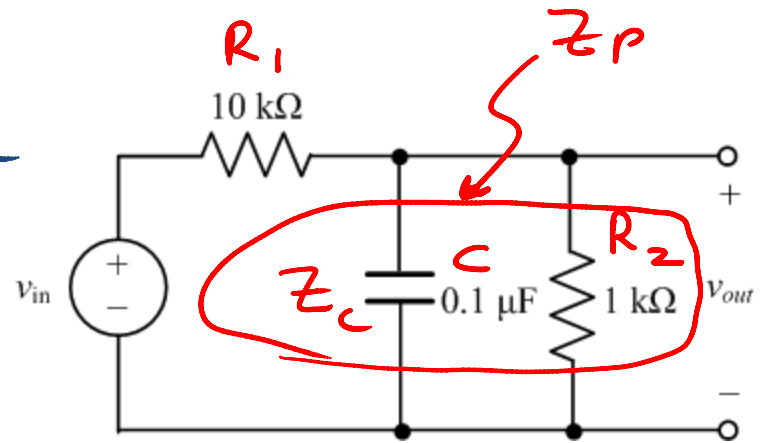
$$Z_c = \frac{1}{j\omega C}, \quad Z_p = R_2 \left( \frac{1}{j\omega C} \right) j\omega C$$

$$\left( R_2 + \frac{1}{j\omega C} \right) j\omega C$$

$$Z_p = \frac{R_2}{1 + j\omega R_2 C}$$

$$V_{out} = \frac{V_{in} Z_p}{R_1 + Z_p} \Rightarrow H(\omega) = \frac{Z_p}{R_1 + Z_p} = \frac{\frac{R_2}{1 + j\omega R_2 C}}{\left( R_1 + \frac{R_2}{1 + j\omega R_2 C} \right) (1 + j\omega R_2 C)}$$

$$H(\omega) = \frac{R_2}{R_1 + j\omega R_1 R_2 C + R_2} = \frac{1000}{11000 + j\omega 10^4 10^3 10^{-7}} = \frac{1000}{11,000 + j\omega}$$



$H(\omega) = 1000/(j\omega + 11000)$ , Lowpass filter

# Wolfram Alpha, add abs( ) for mag

www.wolframalpha.com/input/?i=plot+abs%281000%2F%28iw%2B11000%29%29%2C+w%2C+0%2C100000

plot abs(1000/(iw+11000)), w, 0,100000



Examples Random

Assuming "0" is referring to math | Use as a decimal number instead

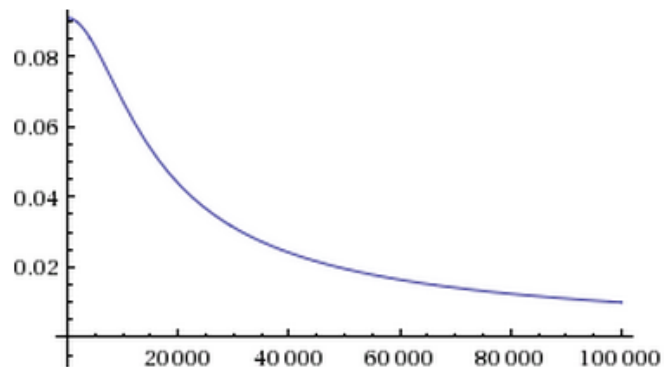
Assuming i is the imaginary unit | Use i as a variable instead

Input interpretation:

plot	$\frac{1000}{ 11000 + iw }$	$w = 0 \text{ to } 100000$
------	-----------------------------	----------------------------

$|z|$  is the absolute value of  $z$  »  
 $i$  is the imaginary unit »

Plot:



Enable interactivity

# Plotting phase in Wolfram

www.wolframalpha.com/input/?i=plot+180%2Fpi\*arg%281000%2F%28iw%2B11000%29%29%2C+w%2C+0%2C100

plot 180/pi\*arg(1000/(iw+11000)), w, 0,100000



Examples Random

Assuming "0" is referring to math | Use as a decimal number instead

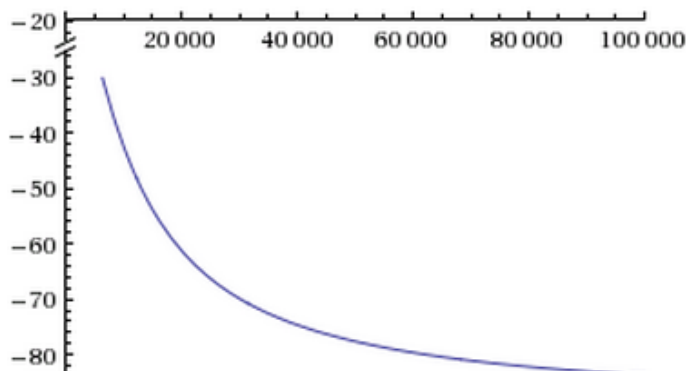
Assuming i is the imaginary unit | Use i as a variable instead

Input interpretation:

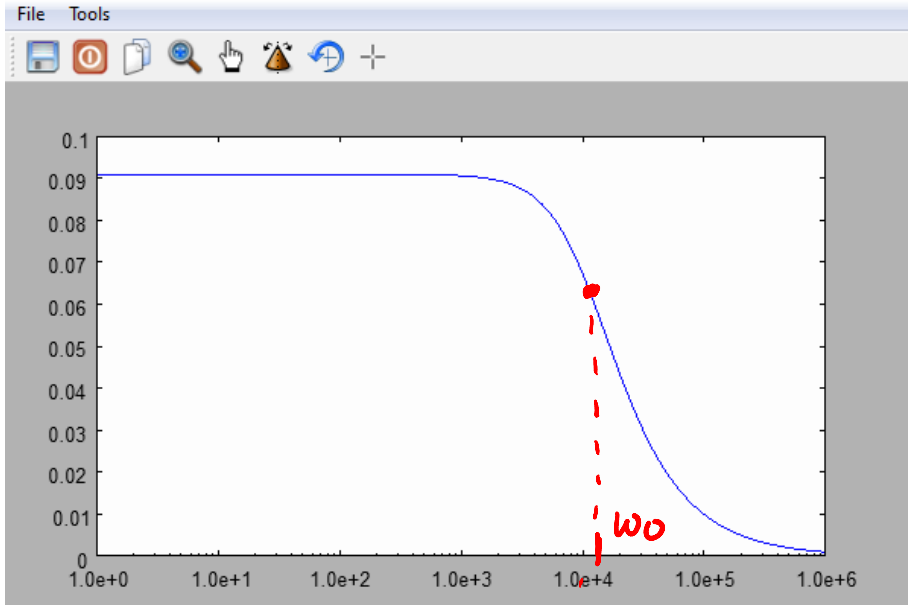
plot	$\frac{180 \operatorname{arg}\left(\frac{1}{11000 + iw}\right)}{\pi}$	$w = 0 \text{ to } 100\,000$
------	---	------------------------------

$\operatorname{arg}(z)$  is the complex argument »  
 $i$  is the imaginary unit »

Plot:



# Magnitude of H(w)



Plot in Matlab:

```
w = [1:100:1e6];
H = 1000./(j*w+11000)
plot(w, abs(H) )
```

semilogx(w,abs(H)) ← magnitude

semilogx(w,angle(H)\*180/pi) ← phase

Find  $\omega_0$  for

$$H(\omega) = \frac{1000}{(11,000 + j\omega)} \cdot \frac{11,000}{11,000}$$

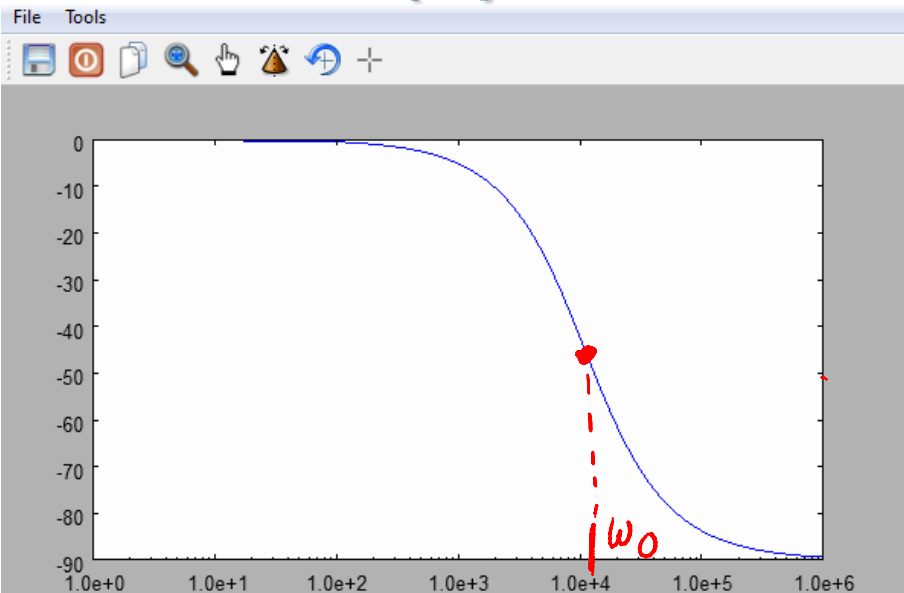
$$H(\omega) = \frac{1/11}{1 + \frac{j\omega}{11000}} = \frac{.0909}{1 + \frac{j\omega}{\omega_0}}$$

$$\omega_0 = 11,000$$

$\omega_0$  is freq at which real and imaginary parts have same magnitude.

$$|H(0)| = .0909 \quad |H(\omega_0)| = \frac{.0909}{\sqrt{2}} = \frac{.064}{14}$$

# Phase of H(w)



Ex 4) Use the Inverting Op-Amp formula to derive the transfer function. Find the cutoff frequency of the filter. Obtain the output voltage in s.s.s. when the input voltage is  $V_{in} = 1\cos(100t)$  V and  $V_{in} = 1\cos(10,000t + 90^\circ)$  V.

Inverting op-amp

$$\text{Gain} = \frac{V_{out}}{V_{in}} = H(\omega) = -\frac{Z_f}{Z_s}$$

$$H(\omega) = \frac{-R}{(R + 1/j\omega C)/R} = \frac{-1}{1 + \frac{1}{j\omega RC}}$$

$$\omega_0 = \frac{1}{RC} = \frac{1}{(10^3)(10^{-7})} = 10^4 \text{ rad/s}$$

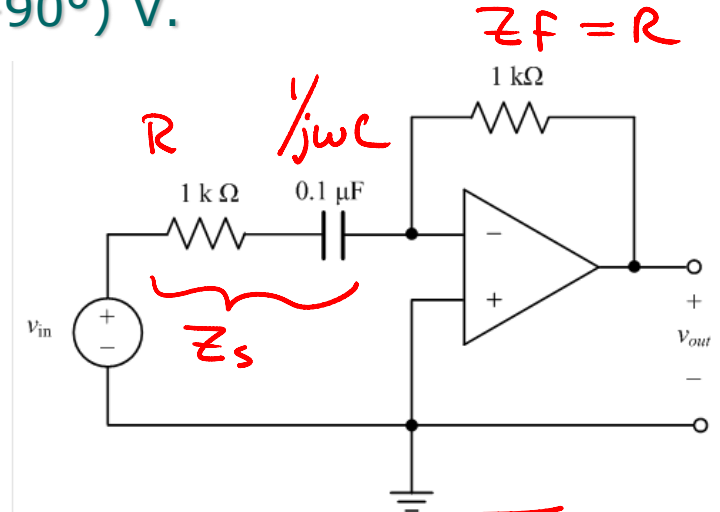
at  $\omega = 100 \text{ rad/s}$

$$V_{in} = 1 \angle 0^\circ \rightarrow V_{out} = V_{in} H(100) = \frac{1 \angle 0^\circ}{j\left(\frac{10^4}{10^2}\right) - 1} = \frac{1}{j100 - 1} \approx \frac{-j}{100}$$

$$\therefore V_{out}(t) = 0.01 \cos(100t - 90^\circ)$$

at  $\omega = 10^4 \text{ rad/s}$   $V_{in} = 1 \angle 90^\circ \rightarrow V_{out} = \frac{1 \angle 90^\circ}{j\left(\frac{10^4}{10^4}\right) - 1} = \frac{1 \angle 90^\circ}{\sqrt{2} \angle 135^\circ} = 0.707 \angle -45^\circ$

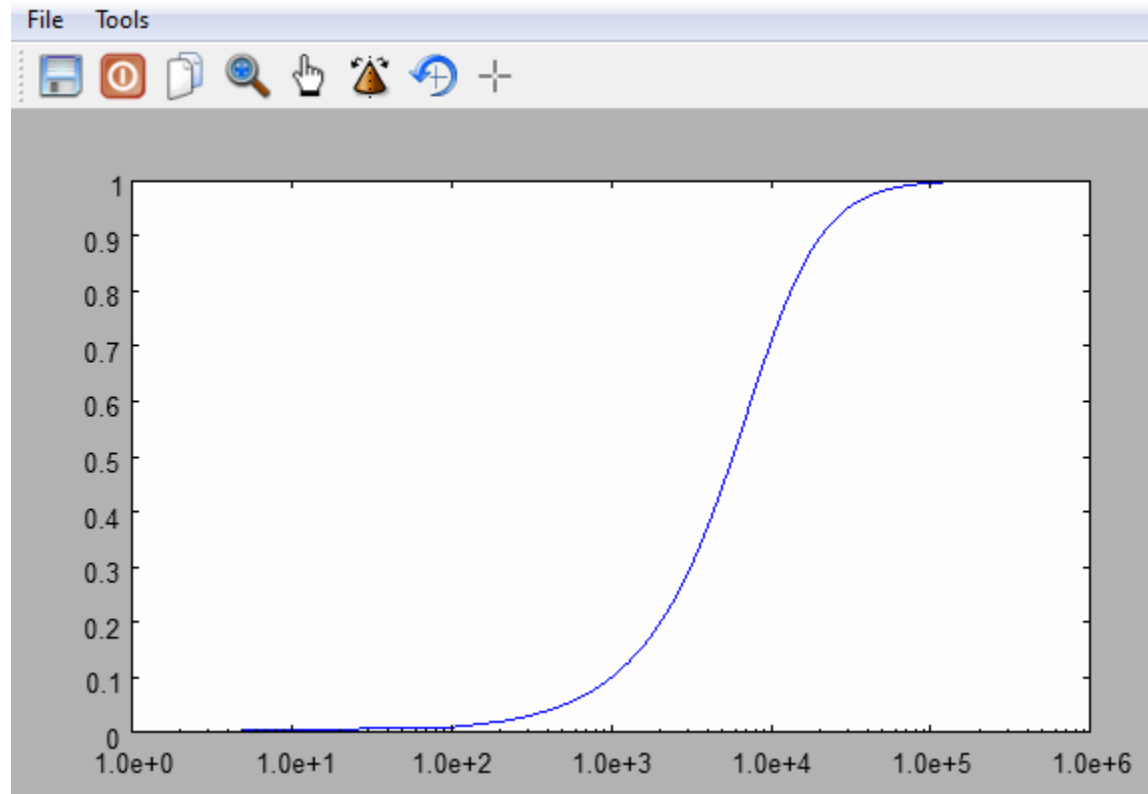
$$\therefore V_{out}(t) = 0.707 \cos(10^4 t - 45^\circ)$$



$$H(\omega) = \frac{1}{j\frac{\omega_0}{\omega} - 1}$$

```
w = [1:100:1e6];  
H = 1./ (j*(10000./w) - 1);  
semilogx(w,abs(H))
```

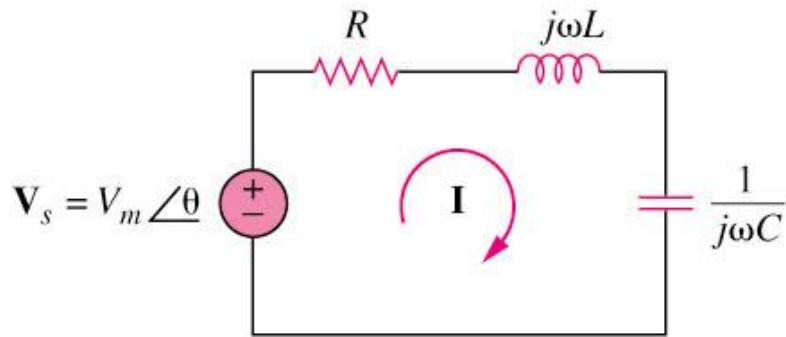
# magnitude





# 14.3 Series Resonance (1)

Resonance is a condition in an RLC circuit in which the capacitive and inductive reactance are equal in magnitude, thereby resulting in purely resistive impedance.



$$Z = R + j\left(\omega L - \frac{1}{\omega C}\right)$$

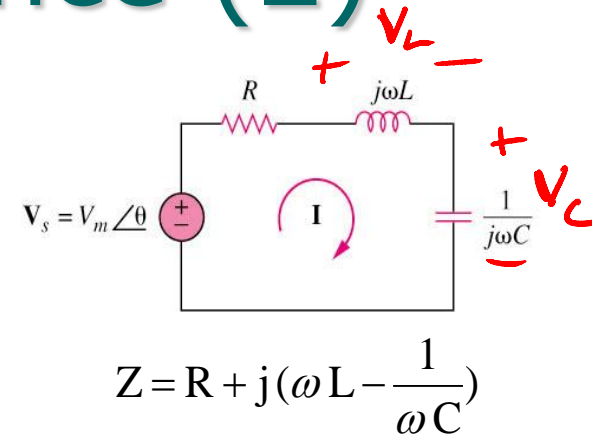
**Resonance frequency:**

$$\text{when } \omega_0 L = \frac{1}{\omega_0 C}$$

$$\text{or } \omega_0 = \frac{1}{\sqrt{LC}}$$

$$f_0 = \frac{1}{2\pi\sqrt{LC}}$$

# 14.3 Series Resonance (2)



## The features of series resonance:

The impedance is purely resistive,  $Z = R$ ;

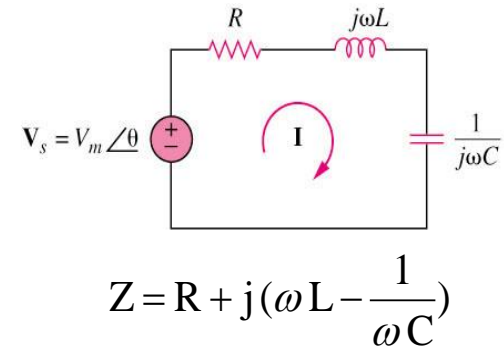
- The supply voltage  $V_s$  and the current  $I$  are in phase, so  $\cos \theta = 1$ ; *(power factor angle = 0)*
- The magnitude of the impedance  $Z(\omega)$  is minimum;
- The inductor voltage and capacitor voltage can be much more than the source voltage.

*180° out of phase so cancel out*

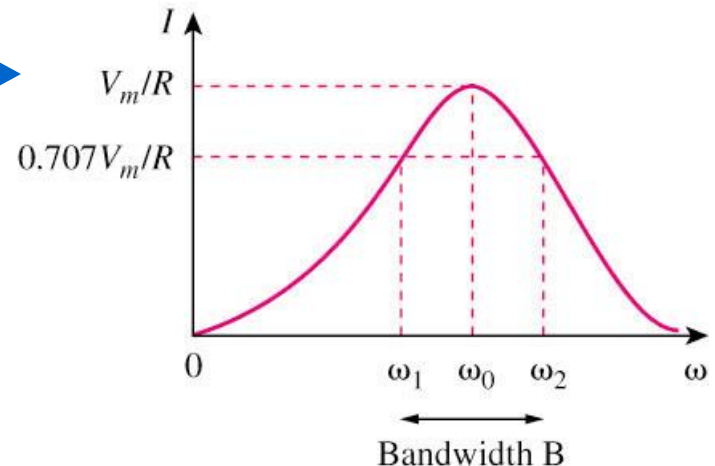
# 14.3 Series Resonance (3)

## Bandwidth B

The frequency response of the resonance circuit current is



$$I = |I| = \frac{V_m}{\sqrt{R^2 + (\omega L - 1/\omega C)^2}}$$



The average power absorbed by the RLC circuit is

$$P(\omega) = \frac{1}{2} I^2 R$$

The highest power dissipated occurs at resonance:

$$P(\omega_o) = \frac{1}{2} \frac{V_m^2}{R}$$

# 14 3 Series Resonance (4)

Half-power frequencies  $\omega_1$  and  $\omega_2$  are frequencies at which the dissipated power is half the maximum value:

$$P(\omega_1) = P(\omega_2) = \frac{1}{2} \frac{(V_m/\sqrt{2})^2}{R} = \frac{V_m^2}{4R}$$

The half-power frequencies can be obtained by setting  $|Z| = R + j(\omega L - \frac{1}{\omega C})$  equal to  $\sqrt{2} R$ :  $\omega L - \frac{1}{\omega C} = R$ ,  $\omega^2 - \frac{R}{L}\omega - \frac{1}{LC} = 0$

$$\omega_1 = -\frac{R}{2L} + \sqrt{\left(\frac{R}{2L}\right)^2 + \frac{1}{LC}}$$

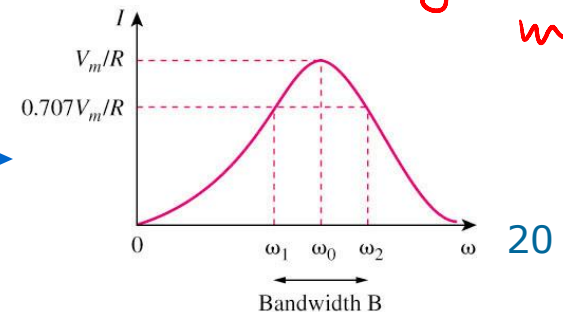
$$\omega_2 = \frac{R}{2L} + \sqrt{\left(\frac{R}{2L}\right)^2 + \frac{1}{LC}}$$

$$\omega_o = \sqrt{\omega_1 \omega_2}$$

*↑ geometric mean*

Bandwidth B

$$B = \omega_2 - \omega_1$$



# 14.3 Series Resonance (5)

Quality factor,

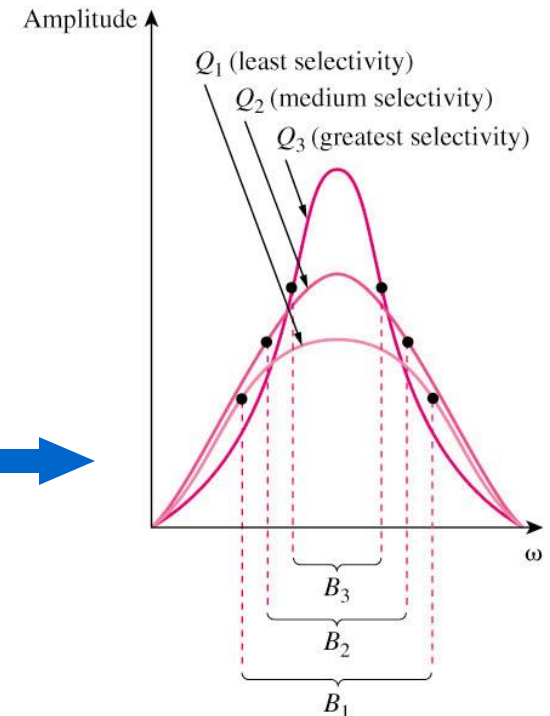
$$Q = \frac{\text{Peak energy stored in the circuit}}{\text{Energy dissipated by the circuit in one period at resonance}} = \frac{\omega_o L}{R} = \frac{1}{\omega_o CR}$$

The relationship between the B, Q and  $\omega_o$ :

$$B = \frac{R}{L} = \frac{\omega_o}{Q} = \omega_o^2 CR$$

$$Q = \frac{\omega_o}{B}$$

- The quality factor is the ratio of its resonant frequency to its bandwidth.
- If the bandwidth is narrow, the quality factor of the resonant circuit must be high.
- If the band of frequencies is wide, the quality factor must be low.



## Example 5

A series-connected circuit has  $R = 4 \Omega$   
and  $L = 25 \text{ mH}$ .

- Calculate the value of  $C$  that will produce a quality factor of 50.
- Find  $\omega_1$  and  $\omega_2$ , and  $B$ .
- Determine the average power dissipated at  $\omega = \omega_0, \omega_1, \omega_2$ .

Take  $V_m = 100\text{V}$ .

$$a) \quad Q = \frac{\omega_0 L}{R} = \frac{1}{\omega_0 CR} = 50 \rightarrow \omega_0 = 50 \left( \frac{R}{L} \right) = 50 \left( \frac{4}{.025} \right) = 8000 \text{ R/s}$$

$$\text{then } C = \frac{1}{Q \omega_0 R} = \frac{1}{(50)(8000)(4)} = \underline{\underline{.625 \mu\text{F}}}$$

$$b) \quad B = \frac{R}{L} = \frac{4}{.025} = 160 \text{ R/s}$$

$$\omega_1 = -\frac{R}{2L} + \sqrt{\left(\frac{R}{2L}\right)^2 + \frac{1}{LC}} = -\frac{4}{.05} + \sqrt{\left(\frac{4}{.05}\right)^2 + \frac{1}{(.025)(.625 \times 10^{-6})}}$$

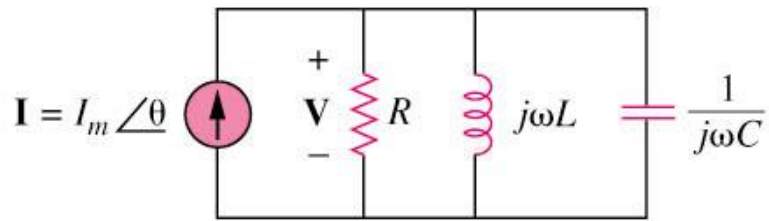
$$= -80 + 8000.4 = \underline{\underline{7920 \text{ R/s}}}, \quad \omega_2 = +80 + 8000.4 = \underline{\underline{8080 \text{ R/s}}}$$

$$c) \quad P(\omega_0) = \frac{1}{2} \frac{V_m^2}{R} = \frac{1}{2} \frac{(100)^2}{4} = 1250 \text{ W}, \quad P(\omega_1) = P(\omega_2) = 625 \text{ W}$$

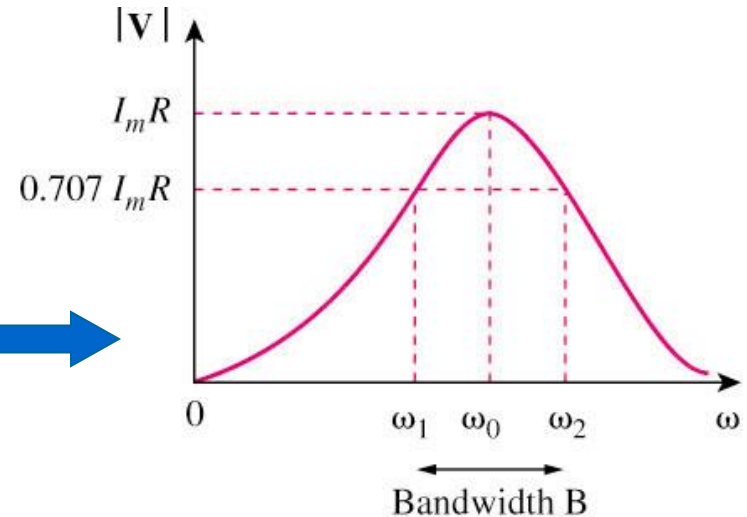


# 14.4 Parallel Resonance (1)

It occurs when imaginary part of Y is zero



$$Y = \frac{1}{R} + j\left(\omega C - \frac{1}{\omega L}\right)$$



**Resonance frequency:**

$$\omega_o = \frac{1}{\sqrt{LC}} \text{ rad/s} \quad \text{or} \quad f_o = \frac{1}{2\pi\sqrt{LC}} \text{ Hz}$$



# 14.4 Parallel Resonance (2)

Summary of series and parallel resonance circuits:

<i>characteristic</i>	<i>Series circuit</i>	<i>Parallel circuit</i>
$\omega_0$	$\frac{1}{\sqrt{LC}}$	$\frac{1}{\sqrt{LC}}$
$Q$	$\frac{\omega_0 L}{R}$ or $\frac{1}{\omega_0 RC}$	$\frac{R}{\omega_0 L}$ or $\omega_0 RC$
$B$	$\frac{\omega_0}{Q}$	$\frac{\omega_0}{Q}$
$\omega_1, \omega_2$	$\omega_0 \sqrt{1 + \left(\frac{1}{2Q}\right)^2} \pm \frac{\omega_0}{2Q}$	$\omega_0 \sqrt{1 + \left(\frac{1}{2Q}\right)^2} \pm \frac{\omega_0}{2Q}$
$Q \geq 10, \omega_1, \omega_2$	$\omega_0 \pm \frac{B}{2}$	$\omega_0 \pm \frac{B}{2}$

## Example 6

(OPTIONAL)

A parallel resonant circuit has  $R = 100 \text{ k}\Omega$ ,  $L = 20 \text{ mH}$ , and  $C = 5 \text{ nF}$ . Calculate  $\omega_0$ ,  $\omega_1$ ,  $\omega_2$ ,  $Q$  and  $B$

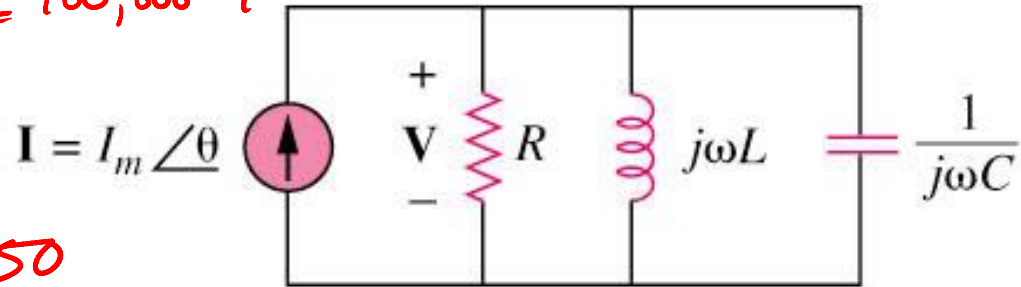
$$\omega_0 = \frac{1}{\sqrt{LC}} = \frac{1}{\sqrt{(0.02)(5 \times 10^{-9})}} = 100,000 \text{ R/s}$$

$$Q = \frac{R}{\omega_0 L} = \frac{100,000}{(100,000)(0.02)} = 50$$

$$\begin{aligned} \omega_1 &= \omega_0 \sqrt{1 + \left(\frac{1}{2Q}\right)^2} - \frac{\omega_0}{2Q} = 100,000 \sqrt{1 + \left(\frac{1}{100}\right)^2} - \frac{100,000}{100} \\ &= 100,000 \sqrt{1.00005} - 1000 \\ &= 99,000 \text{ R/s} \end{aligned}$$

$$\omega_2 = \text{_____} + \frac{\omega_0}{2Q} = 101,000 \text{ R/s}$$

$$B = \omega_2 - \omega_1 = 2000 \text{ R/s}$$



# 14.5 Passive Filters (1)

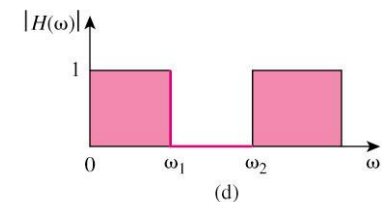
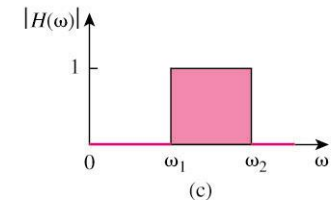
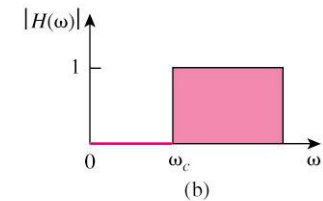
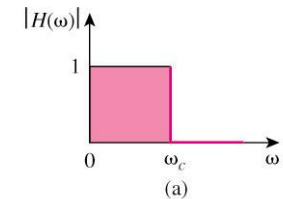
- A filter is a circuit that is designed to pass signals with desired frequencies and reject or attenuate others.
- **Passive filter** consists of only passive element R, L and C.
- There are four types of filters.

**Low Pass**

**High Pass**

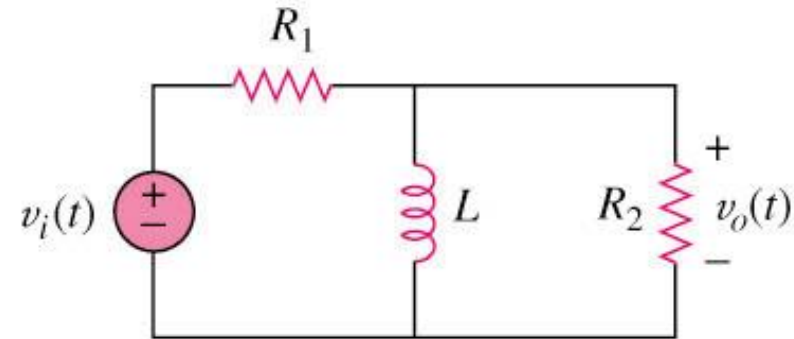
**Band Pass**

**Band Stop**



# Example 7

For the circuit in the figure below, obtain the transfer function  $V_o(\omega)/V_i(\omega)$ . Identify the type of filter the circuit represents and determine the corner frequency. Take  $R_1=100\Omega =R_2$  and  $L =2\text{mH}$ .



$$H(\omega) = \frac{V_o}{V_i} = \frac{Z_L \parallel R_2}{R_1 + Z_L \parallel R_2}$$

$$= \frac{j\omega L R_2 (j\omega L + R_2)}{j\omega L + R_2}$$

$$\left( R_1 + \frac{j\omega L R_2}{j\omega L + R_2} \right) (j\omega L + R_2) = \frac{j\omega L R_2 / (j\omega L)}{(j\omega L R_1 + R_1 R_2 + j\omega L R_2) / j\omega L}$$

$$= \frac{R_2}{R_1 + R_2 + \frac{R_1 R_2}{j\omega L}} = \frac{R_2}{R_1 + R_2 - \frac{j R_1 R_2}{\omega L}}$$

$$H(0) = 0, \quad H(\infty) = \frac{R_2}{R_1 + R_2} \rightarrow \text{High Pass FILTER}$$

$$\text{Corner Freq when } R_1 + R_2 = \frac{R_1 R_2}{\omega_c L} \rightarrow \omega_c = \frac{R_1 R_2}{(R_1 + R_2) L} = \frac{(100)(100)}{(200)(.002)} = 25,000 \text{ rad/s}$$

Answer:  $\omega = 25 \text{ krad/s}$

EXAMPLE 8) a) Find the transfer function  $H(\omega) = V_{out}/V_{in}$

b) At what frequency will the magnitude of  $H(\omega)$  be maximum and what is the maximum value of the magnitude of  $H(\omega)$ ?

c) At what frequency will the magnitude of  $H(\omega)$  be minimum and what is the minimum value of the magnitude of  $H(\omega)$ ?

$$H(\omega) = \frac{V_{out}}{V_{in}} = -\frac{Z_F}{Z_S} = -\frac{R_2 \parallel \frac{1}{j\omega C}}{R_1}$$

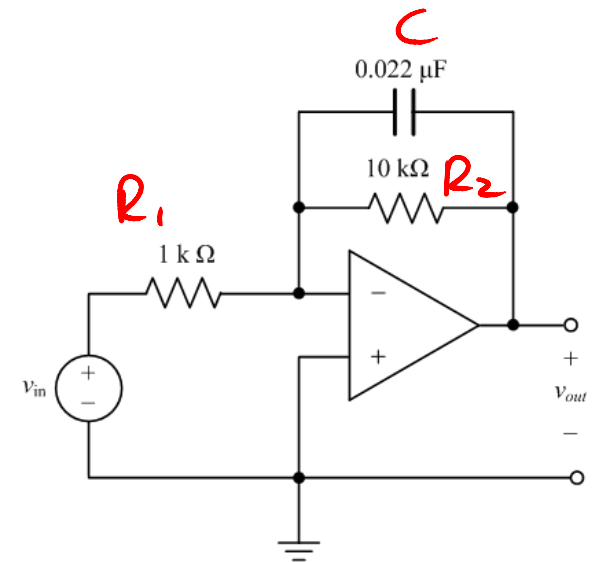
$$H(\omega) = -\frac{\left(\frac{R_2}{j\omega C}\right)}{\left(R_2 + \frac{1}{j\omega C}\right)} \cdot \frac{1}{R_1} = -\frac{1}{R_1} \left(\frac{R_2}{1 + j\omega R_2 C}\right)$$

$$= -\frac{R_2}{R_1 + j\omega R_1 R_2 C} = -\frac{R_2 / R_1 R_2 C}{j\omega + \frac{R_1}{R_1 R_2 C}} = -\frac{1}{(1000)(0.022 \times 10^{-6})} \cdot \frac{1}{j\omega + \frac{1}{(10000)(0.022 \times 10^{-6})}}$$

$$= -\frac{45455}{j\omega + 4545.5}$$

$$H_{min} = 0 \text{ when } \omega \rightarrow \infty$$

$$H_{max} = 10 \text{ when } \omega \rightarrow 0$$



a)  $H(\omega) = -45455/(j\omega + 4545.5)$

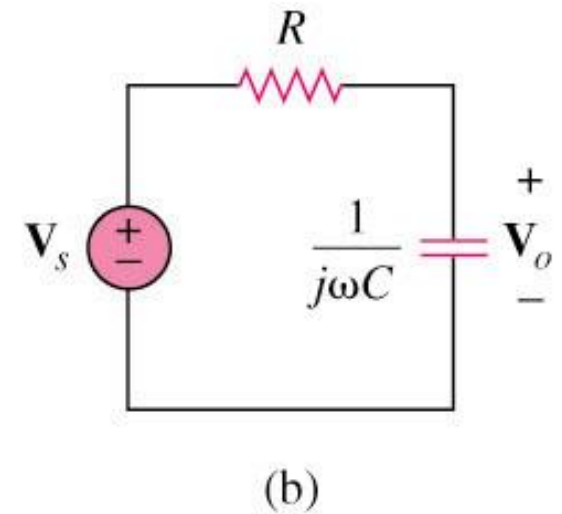
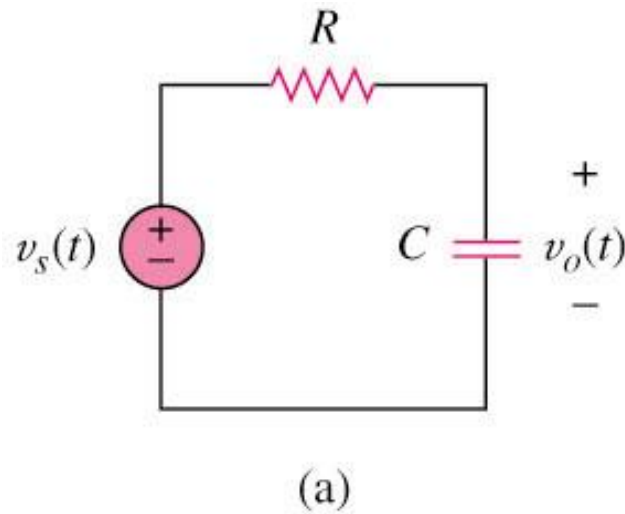
b)  $H_{max} = 10$  as  $\omega = 0$  rad/s

c)  $H_{min} = 0$  as  $\omega$  goes to infinity.

# HANDOUTS

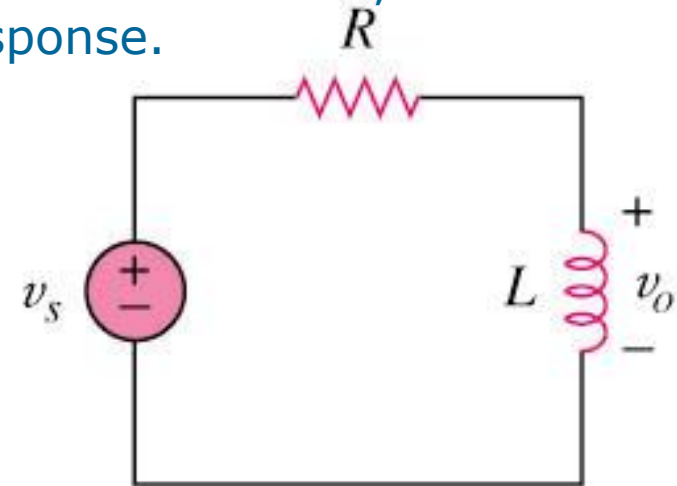
# Example 1

For the RC circuit shown below, obtain the transfer function  $V_o/V_s$  and its frequency response. Let  $v_s = V_m \cos \omega t$ .



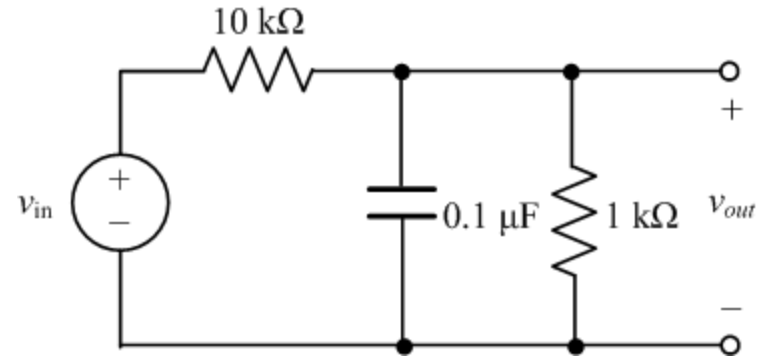
## Example 2

Obtain the transfer function  $V_o/V_s$  of the RL circuit shown below, assuming  $v_s = V_m \cos \omega t$ . Sketch its frequency response.



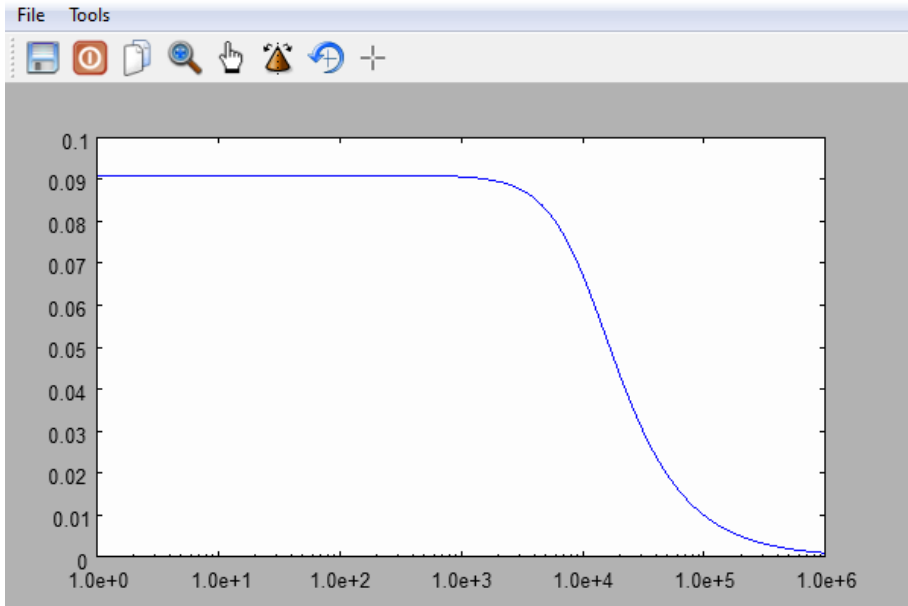


Ex3) Find the transfer function  $H(\omega) = V_{out}/V_{in}$ .  
What type of filter is it?



$H(\omega) = 1000/(j\omega + 11000)$ , Lowpass filter

# Magnitude of H(w)



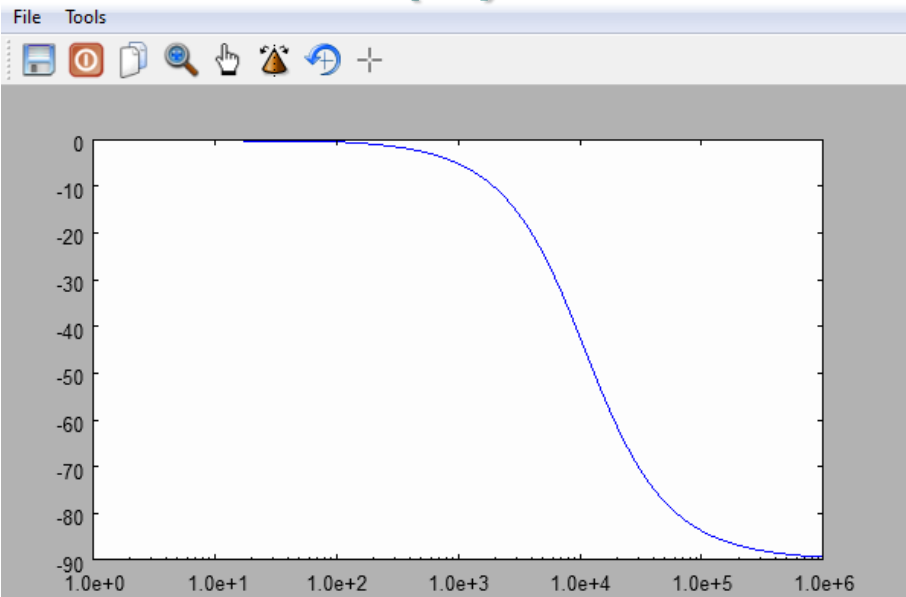
Plot in Matlab:

```
w = [1:100:1e6];  
H = 1000./(j*w+11000)  
plot(w, abs(H) )
```

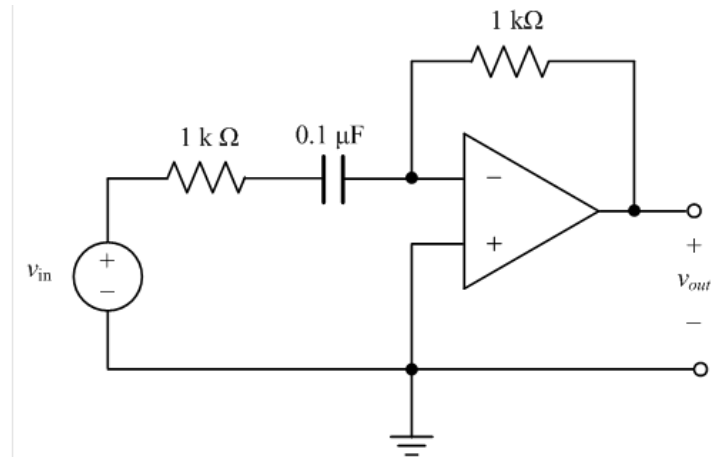
`semilogx(w,abs(H))` ← magnitude

`semilogx(w,angle(H)*180/pi)` ← phase

# Phase of H(w)



Ex 4) Find the cutoff frequency of the filter. Use voltage divider to derive the transfer function. Obtain the output voltage in s.s.s when the input voltage is  $V_{in}=1\cos(100t)$  V and  $V_{in}=1\cos(10,000t+90^\circ)$  V.



## **Example 5**

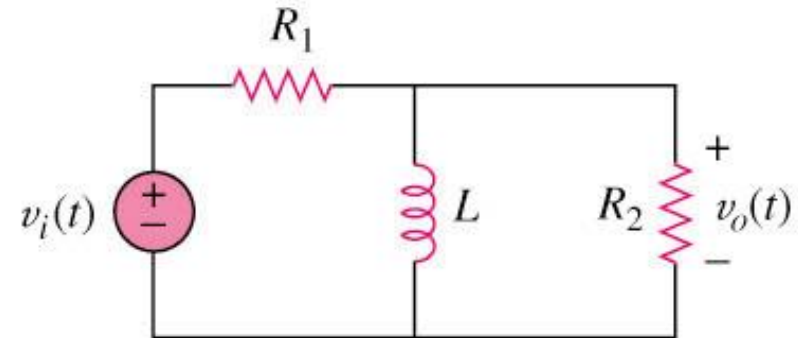
A series-connected circuit has  $R = 4 \Omega$   
and  $L = 25 \text{ mH}$ .

- a. Calculate the value of  $C$  that will produce a quality factor of 50.
- b. Find  $\omega_1$  and  $\omega_2$ , and  $B$ .
- c. Determine the average power dissipated at  $\omega = \omega_0, \omega_1, \omega_2$ .

Take  $V_m = 100\text{V}$ .

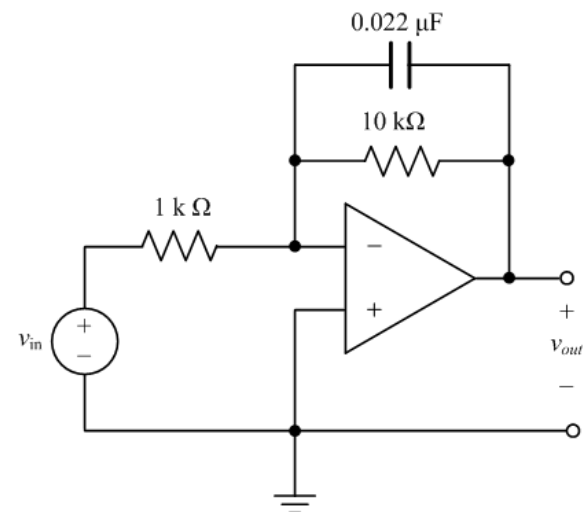
## Example 7

For the circuit in the figure below, obtain the transfer function  $V_o(\omega)/V_i(\omega)$ . Identify the type of filter the circuit represents and determine the corner frequency. Take  $R_1=100\Omega =R_2$  and  $L =2\text{mH}$ .



Answer:  $\omega = 25 \text{ krad/s}$

- EXAMPLE 8) a) Find the transfer function  $H(\omega)=V_{out}/V_{in}$   
 b) At what frequency will the magnitude of  $H(\omega)$  be maximum and what is the maximum value of the magnitude of  $H(\omega)$ ?  
 c) At what frequency will the magnitude of  $H(\omega)$  be minimum and what is the minimum value of the magnitude of  $H(\omega)$ ?



- a)  $H(\omega) = -45455/(j\omega + 4545.5)$   
 b)  $H_{max} = 10$  as  $\omega = 0$  rad/s  
 c)  $H_{min} = 0$  as  $\omega$  goes to infinity.